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FAST COMPUTER ANALYSIS OF NONLINEAR SYSTEMS

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ABSTRACT

Computer aided design of nonlinear differential systems require many response calculations from the system model $\dot{x} = F(x,u)$. Many nonlinear systems, especially control systems, can be separated into a stiff linear part and a nonstiff nonlinear part. Fast numerical integration techniques taking advantage of the partitioned form are presented.

I. INTRODUCTION

Computer aided design of nonlinear differential systems requires many response calculations from the system model $\dot{x} = F(x,u)$ where x is the state vector and u is the control vector. Accurate models include both short and long term effects, and thus have widely separated or stiff eigenvalues. For large stiff sets of equations the analysis takes excessive execution time as most integration techniques (including the Runge Kutta rules) require that the integration time step be limited by the magnitude of the largest and possibly most uninteresting eigenvalues.

Many nonlinear differential systems, especially control systems, can be separated into a linear and a nonlinear part of the form $\dot{x} = Ax + f(x,u)$ where all the stiff eigenvalues are in the linear part represented by the matrix A. This partitioning is possible for most cases because the nonlinear equations usually describe the motion of the system to be controlled (e.g. airplane, missile, hydrofoil, etc.) which has time responses much slower than those of the controlling sensors, compensators and actuators. Further, most sensors, compensators and fast actuators are designed to operate in a linear or almost linear region of their response curves. Nonlinearities, such as physical limits on actuator travel, can most often be incorporated into the plant equation as input saturation. Thus the total system including feedback which is represented by

F(x,u) can be partitioned so that the large eigenvalues occur in the linear part.

In the numerical response calculation for nonlinear differential systems, there are three types of errors to consider; truncation error, roundoff error, and error due to numerical instability of the integrating rule. Roundoff error can be controlled by use of sufficient word length and programming techniques such as inner-product accumulation. Truncation error is the one step error due to the discrete approximation to the integral. This error is a function of the integration rule, the time step size, and the state of the system. For a particular integration rule and step size, bounds for the truncation error can be determined, however, actual truncation error can be significantly less depending on the state of the system. For example, if the state of the system is a stable singular point of the nonlinear equations, truncation error could be negligible. The third type of error, that arising from numerical instability of the integration rule, is often dismissed because truncation error control schemes will normally force the step size small enough so that the integration rule is stable. For stiff systems this may cause an unreasonably large number of integration steps to be required by some rules.

Numerical stability of integration rules [1,2] normally depends on the product of the maximum modulus eigenvalue (of the Jacobian) and the time step size T. For example, for a Runge Kutta fourth order rule $|\lambda| = 1$ must be less than 2.8. Thus even though the truncation error might be negligible for larger step sizes. Thus the remain small to guarantee the computed solution does not grow without bound.

The purpose of this paper is to present several integration rules

which are stable over much larger regions than standard rules and thus allow the step size to be increased when the truncation error allows. For these rules, the region of numerical stability depends on the maximum modulus eigenvalue of the nonlinear part only, which due to partitioning may be orders of magnitude less than that of the composit system.

II. RESPONSE CALCULATION

Transformation

Assume a system of the form $\dot{x} = Ax + f(x,u)$ where stiff eigenvalues are only in the linear part A. Transforming the state variables, $x = e^{At}y$, gives a new set of nonstiff differential equations, $\dot{y} = G(y,u)$ obtained as follows,

$$\dot{x} = Ax + f(x, u) \tag{1}$$

$$Ae^{At}y + e^{At}\dot{y} = Ae^{At}y + f(e^{At}y,u)$$
 (2)

$$\dot{y} = e^{-At} f(e^{At} y, u) .$$
(3)

The dynamic eigenvalues of a nonlinear system are the eigenvalues of the instantaneous Jacobian. The eigenvalues of the original system are the eigenvalues of $A + f_X(x,u)$. The eigenvalues of the transformed system are the eigenvalues of

$$G_y = e^{-At} f_y = e^{-At} f_x x_y = e^{-At} f_x (x, u) e^{At}$$
 (4)

Since G_y is a similarity transformation of f_x , the eigenvalues of the transformed system, G_y are the eigenvalues of the nonlinear part

of the original system, f.

The transformed system of equations can be integrated by fast explicit integration techniques without requiring small time steps to prevent numerical instabilities due to stiff eigenvalues. Integrating \dot{y} over one time step, $t_{n+1} - t_n = T$, gives

$$y_{n+1} = y_n + \int_{t_n}^{t_{n+1}} G(y, u) dt$$
 (5)

Premultiplying by $e^{At_{n+1}}$ gives the convolution integral sequential difference approximation to x_{n+1} .

$$x_{n+1} = e^{AT}x_n + \int_{t_n}^{t_{n+1}} e^{A(t_{n+1}-t)} f(x,u)dt$$
 (6)

Numerical stability is maintained by using either an exact analytic expression for e^{AT} or by using the numerically stable Pade approximations. Table 1 presents some symmetric Pade approximations to e^{AT} . Numerical techniques for the analytical determination of e^{AT} using similarity transformations are similar to those presented in [3].

Table 1. Numerically stable symmetric Pade approximations to e^{AT} .

Order

2
$$(I - \frac{1}{2}TA)^{-1} (I + \frac{1}{2}TA)$$

4 $(I - \frac{1}{2}TA + \frac{1}{12}T^{2}A^{2})^{-1} (I + \frac{1}{2}TA + \frac{1}{12}T^{2}A^{2})$
6 $(I - \frac{1}{2}TA + \frac{1}{10}T^{2}A^{2} - \frac{1}{120}T^{3}A^{3})^{-1} (I + \frac{1}{2}TA + \frac{1}{10}T^{2}A^{2} + \frac{1}{120}T^{3}A^{3})$
8 $(I - \frac{1}{2}TA + \frac{3}{28}T^{2}A^{2} - \frac{1}{84}T^{3}A^{3} + \frac{1}{1680}T^{4}A^{4})^{-1}(I + \frac{1}{2}TA + \frac{3}{28}T^{2}A^{2} + \frac{1}{1680}T^{4}A^{4})$

The above transformed system of equations can be stably integrated by any explicit integration rule such as the Adams-Basforth multistep rules. The Runge Kutta rules are used here because they are self starting and allow easier error detection and correction.

Transformed Runge Kutta rules for order's 1, 2 and 4 are as follows:

1
$$x_{n+1} = e^{AT}[x_n + T f(x_n)]$$
 (7)

2
$$k = Te^{AT}f(x_n)$$
 (8)
 $x_{n+1} = e^{AT}x_n + \frac{1}{2}T\left[e^{AT}f(x_n) + f(e^{AT}x_n + k)\right]$

$$4 k_1 = T f(x_n) (9)$$

$$k_2 = T f(e^{\frac{1}{2}AT}x_n + \frac{1}{2}e^{\frac{1}{2}AT}k_1)$$

$$k_3 = T f(e^{\frac{1}{2}AT}x_n + \frac{1}{2}k_2)$$

$$k_4 = T f(e^{AT}x_n + e^{\frac{1}{2}AT}k_3)$$

$$x_{n+1} = e^{AT}(x_n + \frac{1}{6}k_1) + e^{\frac{1}{2}AT}\frac{1}{3}(k_2 + k_3) + \frac{1}{6}k_4$$

Truncation Error

An example stiff system of equations in a later section of this paper demonstrates how the transformed rules maintain numerical stability for large step sizes even though the truncation errors become large. For a linear system of equations, $\dot{x} = Cx$, the truncation error for a p-th order Runge Kutta rule is

$$e_n = x(t_n) - x_n = \frac{T^{p+1}c^{p+1}}{(p+1)!} x_n + O(T^{p+2})$$
 (10)

For a transformed Runge-Kutta rule applied to $\dot{x} = Ax + f - if - f - Rx$.

then the truncation error for a p-th order rule using a p-th order Pade approximation is dominated by

$$e_n = KT^{p+1}A^{p+1}x_n \tag{11}$$

since ||A|| >> ||B||. If the approximation to e^{AT} is of order greater than p then

$$\mathbf{e}_{\mathbf{n}} = \mathbf{K} \mathbf{T}^{\mathbf{p+1}} \mathbf{A}^{\mathbf{p}} \mathbf{B} \mathbf{x}_{\mathbf{n}} . \tag{12}$$

For T much greater than $||A||^{-1}$ the error gets large so that significant increases in T cannot be made.

For example if A has eigenvalues like -1000 and f_X has eigenvalues like -1 then a standard Runge Kutta rule will require a time step of about .001 to maintain numerical stability. The transformed method will be numerically stable for any step size up to 1, but the truncation error can be like $T^{p+1}10^{3p}/(p+1)$!. To control this truncation error T might still have to be kept small.

Convolution Integral Partitioning

When the stiff system of equations can be further partitioned so that part of the system has no stiff part as is shown in figure 1 and equation 13, then low error convolution integral approximations can be made.

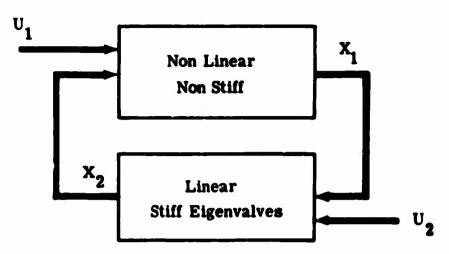


Figure 1. DIAGRAM OF PARTITIONED SYSTEM

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} - \begin{bmatrix} 0 & 0 \\ 0 & A_2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} f_1(x_1, x_2, u_1) \\ f_2(x_1, x_2, u_2) \end{bmatrix}$$
(13)

Numerically stable, low error integration rules can be obtained by applying first and second order explicit one step end point approximations to the convolution integral representation of \mathbf{x}_2 .

$$x_2^{n+1} = e^{A_2 T} x_2^n + \int_{t_n}^{t_{n+1}} e^{A_2(t_{n+1}-t)} f_2(t) dt$$
 (14)

This approximation to \mathbf{x}_2 is coupled with compatible Runge Kutta rules to approximate \mathbf{x}_1 .

A first order approximation to the convolution integral can be obtained by expanding f_2 about t_n and then integrating the exponential term exactly.

$$f_2(x(t)) = f_2^n + O(T)$$
 (15)

$$\int_{t_n}^{t_{n+1}} e^{A_2(t_{n+1}-t)} dt = h_2^{-1} (e^{A_2T} - 1) = I_1$$
 (16)

The integral I₁ is well defined even if A₂ is not of full rank and can be calculated, as before, using stable Pade approximations or evaluating exact analytical expressions. Using equations (15) and (16), equation (14) becomes

$$x_2^{n+1} = e^{A_2T}x_2^n + I_1f_2^n + O(T^2)$$
 (17)

From a first order rule

$$x_1^{n+1} = x_1^n + Tf_1^n \tag{18}$$

Similarly to get a second order rule

$$f_2(x(t)) = f_2^n + f_2^n [t-t_n] + O(T^2)$$
 (19)

$$\int_{t_n}^{t_{n+1}} e^{A_2(t_{n+1}-t)} [t-t_n] dt = A_2^{-1}(t_1 - TI) = I_2$$
 (20)

A first order approximation to \hat{f}_2^n is needed so evaluating f_2 at x_1^{n+1} and x_2^{n+1} from (17) and (18) gives a first order approximation to f_2^{n+1} .

$$f_2^n = f_2^{n+1} - f_2^n + o(r^2)$$
 (21)

Using equations (19), (20), and (21) equation (14) becomes

$$x_2^{n+1} = e^{A_2T}x_2^n + i_1f_2^n + i_2[f_2^{n+1} - f_2^n]/T + O(T^3)$$
 (22)

From a second order Runge Kutta rule

$$x_1^{n+1} = x_1^n + \frac{1}{2}T(f_1^n + f_1^{n+1})$$
 (23)

where again f_1^{n+1} is evaluated at the first order points in equations (17) and (18). Equations (22) and (23) define a partitioned second order rule.

Under the special circumstances when f_2 is a function of only x_1 and the Jacobian of f_2 is readily available (such as when f_2 is linear) then slightly better accuracy can be obtained by using the following approximation to f_2 :

$$f_2(x_1(t)) = f_2^n + f_{2x}^n f_1^n [t-t_n] + 0(T^2)$$
 (24)

Now equation (22) takes the form

$$x_2^{n+1} = e^{A_2 T} x_2^n + I_1 f_2^n + I_2 f_{2x}^n f_1^n$$
 (25)

Using (18) and (25) f_1^{n+1} can be approximated to be used to define x_1^{n+1} in equation (23). Equations (22) and (23) define a partitioned second order rule when the Jacobian of f_2 is available.

The truncation errors for these partitioned rules are on the order of the truncation errors for a system of equations with eigenvalues on the order of the eigenvalues of the nonlinear parts. For example if A_2 has eigenvalues like -1000 and f_1 and f_2 have eigenvalues like -1 then the truncation error will be like $T^{p+1}/(p+1)!$. To get 3 place accuracy with a first order rule T should be about .04 and with a second order rule T should be about .1.

III. EXAMPLE

In order to illustrate the application of these solution techniques, consider a simple pendulum problem in which the angular

rate is measured and used as a signal to a torque motor to provide damping. The nonlinear equations describing the motion of the pendulum are

$$ML \ddot{\theta}(t) = -Mg \sin \theta(t) + T(t)/L \qquad (26)$$

where M is the mass, L is the length, and T(t) is the torque applied by the torque motor. Let the dynamics of the angle sensor be given by a first order lag with a time constant $1/\tau_g$. Likewise let the torque motor be represented by a first order lag with time constant $1/\tau_g$. To obtain the state equations, let x_1 be θ , x_2 be $\dot{\theta}$, x_3 be the output of the angular rate sensor, and let x_4 be the torque. The equations of motion are then

$$\begin{bmatrix} \mathbf{x}_{1} \\ \mathbf{x}_{2} \\ \mathbf{x}_{3} \\ \mathbf{x}_{4} \end{bmatrix} = \begin{bmatrix} \mathbf{x}_{2} \\ -\mathbf{g}/\mathbf{L} \sin \mathbf{x}_{1} - \mathbf{x}_{4}/\mathbf{ML}^{2} \\ \mathbf{K}_{\mathbf{s}} \mathbf{x}_{2} - \mathbf{x}_{\mathbf{s}} \mathbf{x}_{3} \\ \mathbf{K}_{\mathbf{m}} \mathbf{m} \mathbf{x}_{3} - \mathbf{x}_{\mathbf{m}} \mathbf{x}_{4} \end{bmatrix}$$
(27)

For the purpose of this example, let g/L be 1., $1/\text{ML}^2 = 0.5$, $\tau_{\rm s} = 1000$., $K_{\rm s} = 1.0$, $\tau_{\rm m} = 50$., and $K_{\rm m} = 3.0$.

Now factor the system into the form

$$\dot{x} = \begin{bmatrix} 0 & 0 & 0 \\ 0 & -1000 & 0 \\ 150 & -50 \end{bmatrix} x + \begin{bmatrix} x_2 \\ \sin x_1 & -.5x_4 \\ 1000 & x_2 \end{bmatrix}$$
 (28)

Note that the Jacobian of F for small x_1 is

$$\frac{\partial \mathbf{F}}{\partial \mathbf{x}} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -1 & 0 & 0 & -.5 \\ 0 & 1000 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}$$
 (29)

which has eigenvalues of 0.0, 0.0, $\pm j1.0$. The eigenvalues of the composit system for small x_1 are -1000.08, -48.371, $-.77484\pm j$.65818.

For the example system of equations considered here some indications of the acceptability of the suggested integration rules can be made. Figure 2 is a plot of discretization error curves for several rules. The log of the error versus the log of the integration time step has a slope which is equal to the order of the integration rule. Table 2 compares the number of significant figures of accuracy obtained for the integrating rules at several time steps. Also computer execution times are given.

A fourth order Runge Kutta rule had high accuracy for small step sizes, but it went unstable for acceptable step sizes. An order two Runge Kutta rule went unstable for smaller step sizes. The transformed Runge Kutta rules remained stable for large step sizes but the accuracy was poor. The partitioned methods gave acceptable accuracy for reasonable step sizes. Using the Jacobian technique gave about 1 place better accuracy than the time derivative approximation technique. Using a Pade approximation to e^{AT} instead of an exact expression did not degrade the accuracy much, but in some problems this may cause loss of accuracy because large negative Pade exponents approach -1. instead of 0. The partitioned 2nd order rule allows integration to acceptable

accuracy about 200 times faster than a standard 4th order Runge Kutta rule for this example.

Table 2. Number of significant figures and timings for example.

Integration rule/time step	10-3	10 ⁻²	10 ⁻¹
Order 4 Runge Kutta	12	unstable	unstable
Order 2 Runge Kutta	unstable	unstable	unstable
Order 2 transformed	3	1	0
Order 2 partitioned	6	4	2
Order 2 partitioned (with Pade approximation)	6	4	2
Order 2 partitioned (with Jacobian)	7	5	3
Execution time in seconds for all order 2 rules (360/44)	30	3	.3

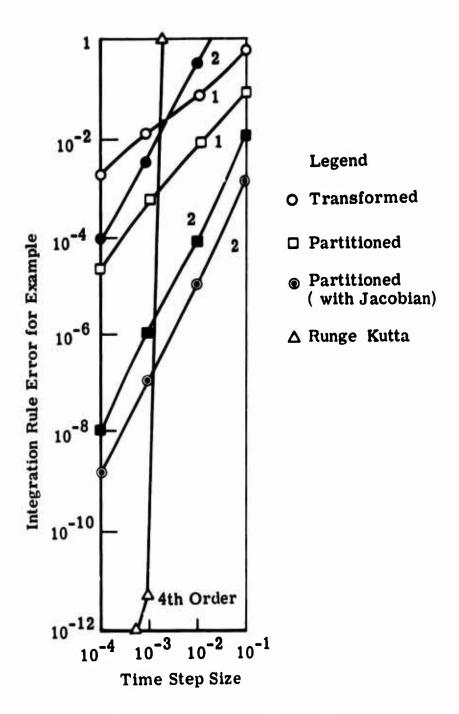


Figure 2. INTEGRATION ERROR CURVES

IV. CONCLUSIONS

In many simulations where one or .1 percent accuracy is sufficient, the larger step sizes allowed by the low order stable integration rules reduces run times appreciably. Control systems for nonlinear plants which use position and rate feedback with compensation usually allow partitioning as indicated in Eq. 13 with f_2 being a function of x_1 and u only. Experience has shown the second order rule to be ideally suited for this problem. Where acceleration feedback is used, f_2 now becomes a function of x_2 in addition to x_1 and u which in some cases reduces the effectiveness of the integration rule.

The rules presented here show advantages over other rules such as standard Runge Kutta only when the truncation errors of the primary variables are small enough to warrant larger step sizes. Where the problem is truncation error limited, the higher order rules offer advantages in error control. It should be noted that in many simulations, however, errors due to numerical instabilities are the limiting factor. For these systems, stable integration rules can significantly reduce computation times.

One obvious example where these rules are particularly effective is in determining steady state operating conditions of nonlinear systems. Here, large errors in the transient solution can be tolerated making stable integration rules with large step sizes quite desirable.

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